#### Projects in Wireless Communication Synchronization

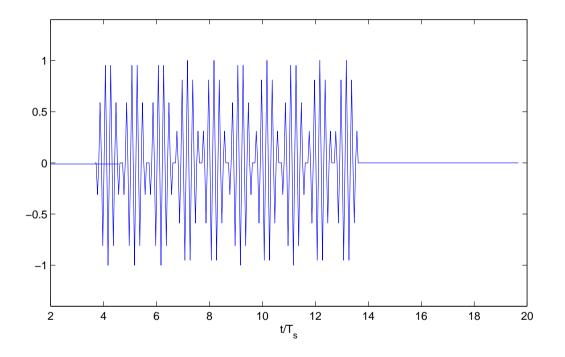
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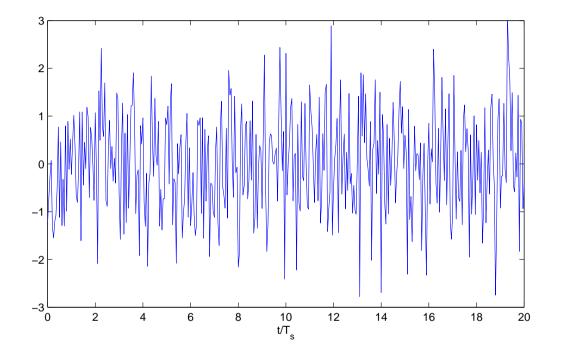
Lund, Fall 2024





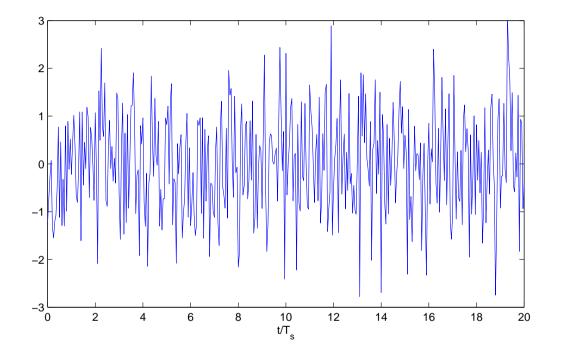
If there is no noise, its easy....





How about this one?





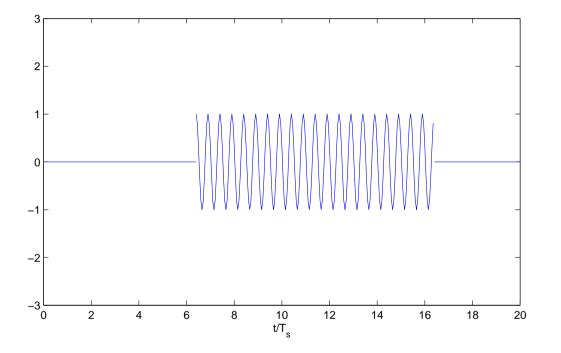
How about this one?

The signal starts at  $t/T_s = 6.5$  and ends at  $t/T_s = 16.5$ .  $E_b/N_0 = 10$  dB.



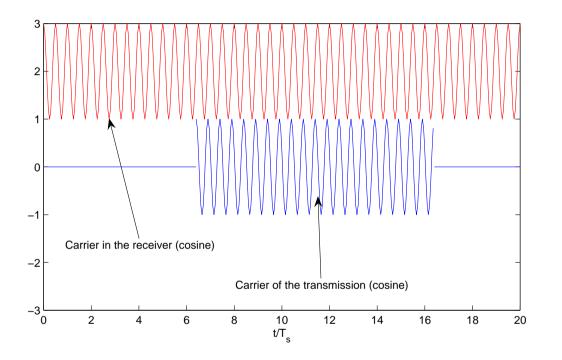
- In general, syncronization is a difficult subject
- Requires solid background in statistical signal processing and control theory
- Phase-locked-loops (PLL) are essential
- We will only make use of simple techniques....





The transmitted signal (in-phase component) contains a cosine

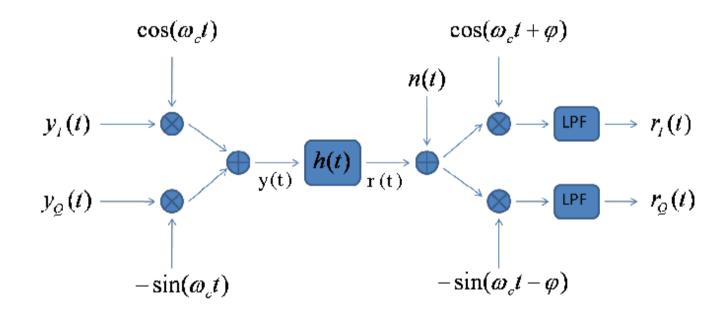




The transmitted signal (in-phase component) contains a cosine If we dont know the exact starting position of the signal, the cosine of the reciever and the cosine of the transmission will be out of phase



The situation can be represented as below.





We derived on the last lecture, that if the transmitted complex symbol is a, the the received symbol will be  $a \exp(-i\phi)$ . Hence, it is crucial that the receiver can estimate  $\phi$  so that it can compensated.



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Moreover, we dont know where the sampling unit (after the matched filter) should sample the signal



## Synchronization

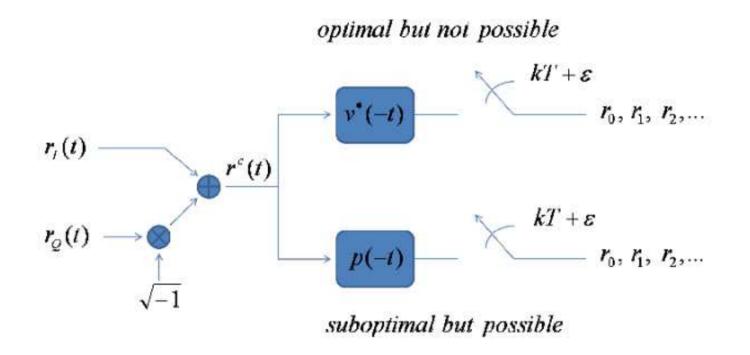
We have that the transmitted signal is based on the pulse shape p(t), and the channel is h(t).

A matched filter should be mathced to the receive pulse  $v(t)=p(t)\star h(t).$  But, this is not possible in cases where the receiver does not know h(t).

Not knowing  $\epsilon$  is severe and will be analyzed next.



The situation can be represented as below.





The phase mismatch and non-optimal sampling instance yields a channel model

 $r[k] = \alpha e^{i\phi} a[k] + n[k]$ 

Phase mismatch gives  $\phi$  and sampling mismatch gives  $\alpha$ 

 $\phi$  can be estimated from a pilot symbol p. Let

$$a[1] = p = 1 + i = \sqrt{2}e^{i\pi/4}$$

then

$$\hat{\phi} = \text{angle}\{r[1]\} - \frac{\pi}{4}$$

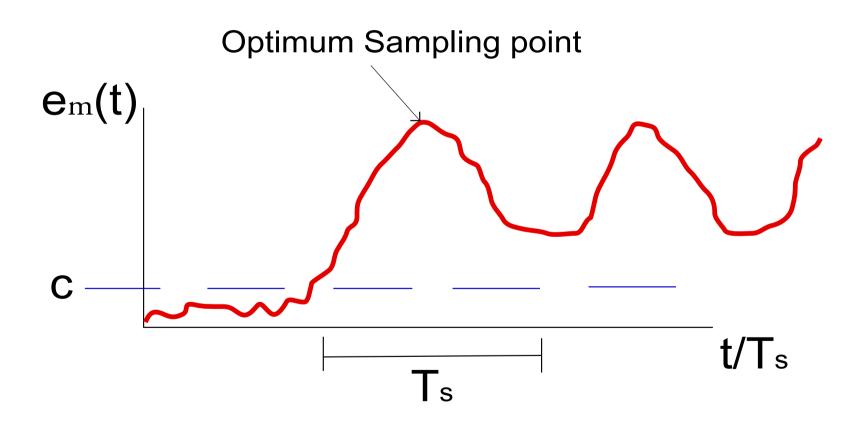


Estimation of  $\epsilon$  is solved by selecting the sampling instance such that  $\alpha$  is maximized. Recall the receiver structure



$$e_m(t) = \sqrt{m_I^2(t) + m_Q^2(t)}$$

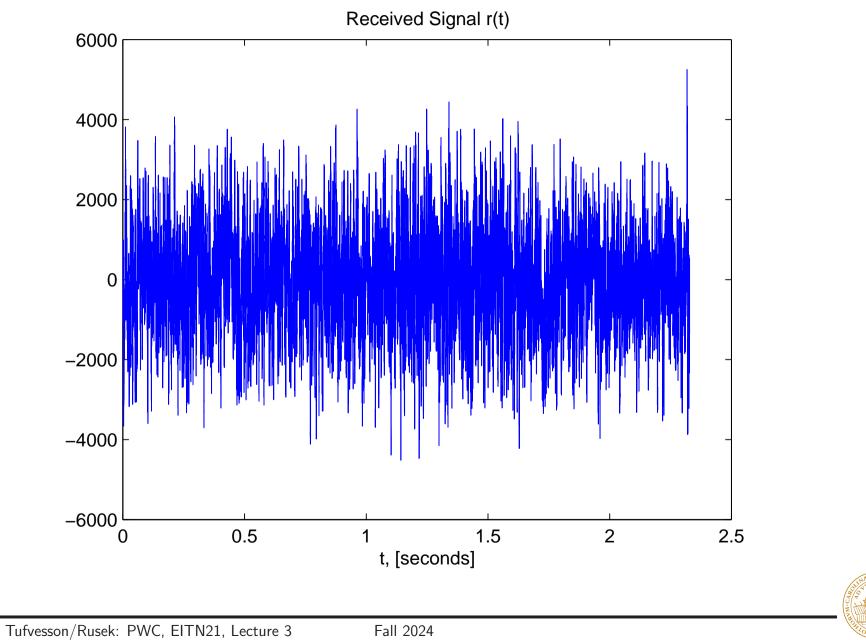
 $\boldsymbol{c}$  is a threshold, easiest to find by trial and error





The transmission starts and ends with pilots 2 + 2i





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Construct  $m_I(t)$  and  $m_Q(t)$  as

$$m_I(t) = r(t)\cos(2\pi f_c t) \star p(t)$$

 $\mathsf{and}$ 

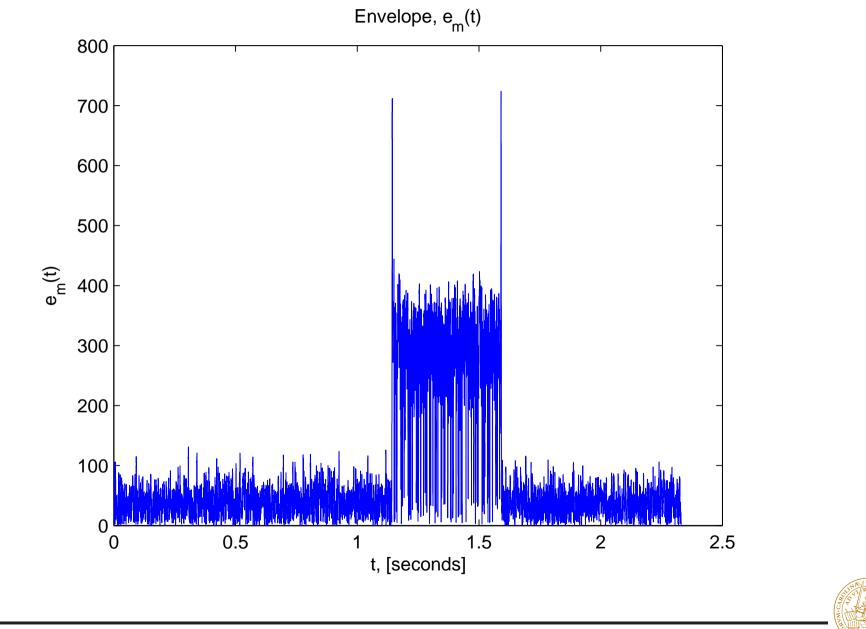
$$m_Q(t) = -r(t)\sin(2\pi f_c t) \star p(t)$$

Then generate

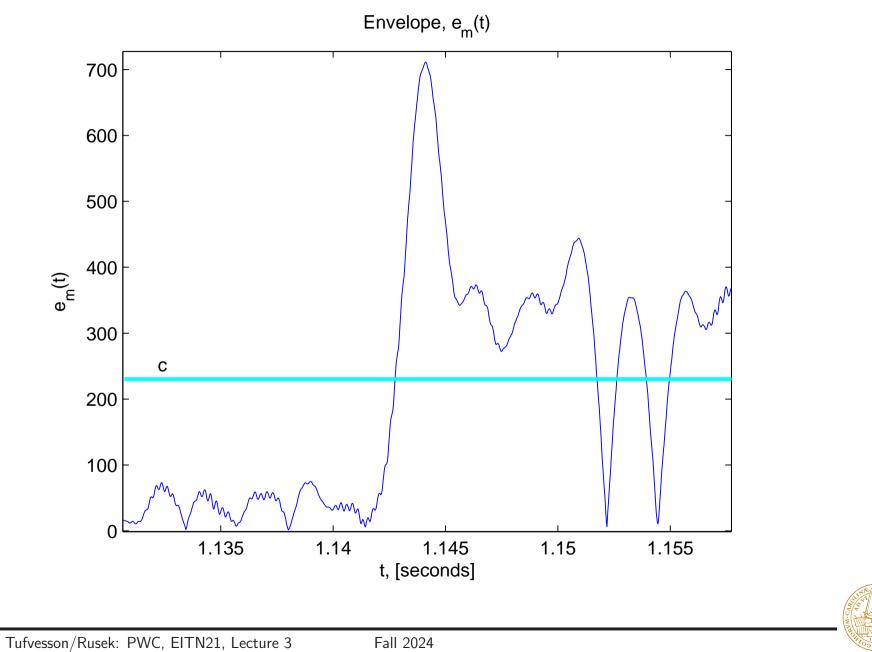
$$e_m(t) = \sqrt{m_I^2(t) + m_Q^2(t)}$$

Plot





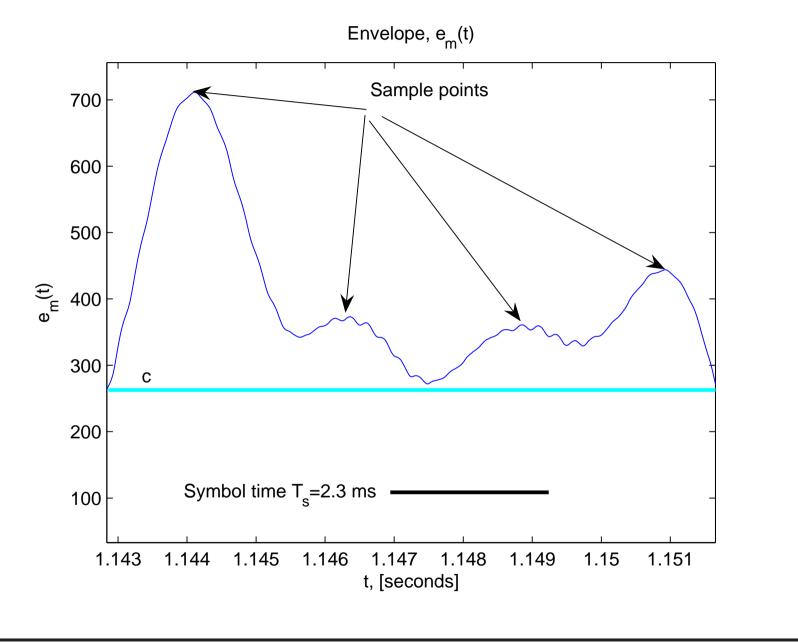
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Optimal sampling point is at  $T_{\text{Sample}} = 1.144$  seconds.





Tufvesson/Rusek: PWC, EITN21, Lecture 3

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Sample at  $T_{\text{Sample}} + kT_s$ :

$$r[k] = m_I(T_{\text{Sample}} + kT_s) + im_Q(T_{\text{Sample}} + kT_s)$$

We get

 $r[k] = 4.38 - 5.60i \ 3.16 + 1.97i \ 2.79 + 2.16i \ 3.55 + 2.66i \ -2.73 - 2.25i...$ Consequently

$$\alpha \exp(i\phi) = r[0]/(2+2i) = -0.30 - 2.49i$$

$$r[1]/\alpha \exp(i\phi) = -0.93 + 1.15i \text{ and } r[2]/\alpha \exp(i\phi) = -0.9899 + 0.9982i$$
  
So  
$$\hat{a}[1] = -1 + i \text{ and } \hat{a}[2] = -1 + i$$

